1. Given that

$$
x(n) = \{4, -3, 2, -1, -5, 3, -1, 0\}, \quad n = 0 \to 7
$$

(a) Using only 2-point FFTs, calculate the FFT of $x(n)$

Heres the MATLAB script. Implemented for any size sequence $x(n)$ where N is a power of 2

```
1 %% 1 part a
2
3 % sequence x(n)
4 \times = [4, -3, 2, -1, -5, 3, -1, 0];5
6 % get N, constant WN
7 \text{ N} = \text{length}(x);8 WN = exp(−1i * 2 * pi / N);
\alpha10 % result of FFT is result x
11 result_x = x;
12 % temporary array holds values, go from one 2−point FFT to another
13 hold_x = bitrevorder(result_x);
14
15 % array to hold exponents for WN
16 exp_array = zeros(1, length(x) / 2);
17 % number of sections for corresponding FFTs
18 blocks = N / 2;
19
20 % loop through each layer of fast DFT algorithm
21 for n = 0:log2(N) - 122
23 % populate array for exponents of WN
24 for k = 0:2ˆn:(N / 2) − 1
25 exp_array(k + 1: k + 2^n) = 0:(N / 2^(n + 1)):(N / 2) - 1;
26 end
27
28 % a = index of exponents for WN
29 a = 1;30 % other algorithm variables, will explain in a figure later
31 idx = 1;32 for m = 1:blocks
33 for idx2 = 0: (2^n n) - 134 index = idx + idx2;
35
36 % do the appropriate 2−point FFT
37 radix_res = fft([hold_x(index), (WN^{\circ}exp\_{array}(a)) \star ...hold_x(index + 2^{\circ}n)], 2);
38
39 % store results into result x
40 result x(index) = radix res(1);
41 result_x(index + 2^nn) = radix_res(2);
4243 a = a + 1;44 end
45 i dx = i dx + 2^(n + 1);
46 end
47
48 blocks = blocks / 2;49 hold_x = result_x;50 end
```


Figure 1: Annotated Radix-2 FFT diagram

Figure courtesy of

<https://riptutorial.com/algorithm/example/27088/radix-2-fft>.

In my algorithm shown above, the whole for loop is represented by the green sections, blocks is represented by the blue sections, the initial bitrevorder (x) is done on x to represent the red section, and the exp_array is represented by the orange sections. As for the indicies, the inner for loop loops through each block to appropriately compute the 2-point DFT.

Here is the resulting plot of result_x in MATLAB

Figure 2: Radix-2 FFT magnitude result

(b) The standard 8-point DFT of $x(n)$ using the MATLAB routine fft looks as such

Figure 3: Standard 8-Point FFT magnitude result

(c) The results are identical, with an absolute difference on the magnitude of 10^{-15}

Figure 4: Absolute difference of results

2. (a) Get $H(z)$ in the form

 $H(z) = E_1(z^4) + z^{-1}E_2(z^4) + z^{-2}E_3(z^4) + z^{-3}E_4(z^4)$

$$
E_1(z) = 0.0168 + 0.0287z^{-1} + 0.2453z^{-2} - 0.0409z^{-3}
$$
 (1)

$$
E_2(z) = 0.0264 - 0.1205z^{-1} - 0.1205z^{-2} + 0.0264z^{-3}
$$
 (2)

$$
E_3(z) = -0.0409 + 0.2453z^{-1} + 0.0287z^{-2} + 0.0168z^{-3}
$$
 (3)

$$
E_4(z) = 0.0334 + 0.6694z^{-1} + 0.0334z^{-2}
$$
\n⁽⁴⁾

(b) Appropriate difference equation is given using the following diagram

$$
x(n) \longrightarrow E_1(z^4) \longrightarrow E_1(z^4) \longrightarrow \bigoplus_{z=1}^{S_1(n)} g_1(n) \longrightarrow E_2(z^4) \longrightarrow \bigoplus_{z=1}^{S_2(n)} g_2(n) \longrightarrow E_3(z^4) \longrightarrow \bigoplus_{z=1}^{S_3(n)} g_3(n) \longrightarrow E_4(z^4) \longrightarrow \bigoplus_{z=1}^{S_4(n)} g_4(n)
$$

$$
g_1(n) = x(n-1) \tag{5}
$$

$$
g_2(n) = g_1(n-1)
$$
 (6)

$$
g_3(n) = g_2(n-1)
$$
 (7)

$$
S_1(n) = 0.0168x(n) + 0.0287x(n-4) + 0.2453x(n-8) - 0.0409x(n-12)
$$
 (8)

$$
S_2(n) = 0.0264g_1(n) - 0.1205g_1(n-4) - 0.1205g_1(n-8) + 0.0264g_1(n-12)
$$
 (9)

$$
S_3(n) = -0.0409g_2(n) + 0.2453g_2(n-4) + 0.0287g_2(n-8) + 0.0168g_2(n-12)
$$
 (10)

$$
S_4(n) = 0.0334g_3(n) + 0.6694g_3(n-4) + 0.0334g_3(n-8)
$$
\n(11)

$$
y(n) = S_1(n) + S_2(n) + S_3(n) + S_4(n)
$$
\n(12)

3. (a) The 2N-point sequence $y(n)$ is defined as

$$
y(n) = \begin{cases} x(\frac{n+1}{2}), & n \text{ odd} \\ 0, & n \text{ even} \end{cases}
$$

where $X(k)$ is the N-point DFT of the sequence $x(n)$ for $n \in \{0, N-1\}$ The 2N-point DFT of $y(n)$ is defined as

$$
Y(k) = \sum_{n=0}^{2N-1} y(n)e^{\frac{-j2\pi nk}{2N}}
$$
\n(13)

And since $\forall n_{\text{even}}$, $y(n) = 0$, the sequence $y(n)$ under the condition that n_{odd} summated equals the summated sequence $x(n)$. The only difference in the DFT would be the *n* in the exponential term, where it ranges from $n = 0 \rightarrow N - 1$ $\forall n_{\text{odd}}$, meaning

$$
\sum_{n=0}^{2N-1} y(n)e^{\frac{-j2\pi nk}{2N}} = \sum_{n=0}^{N-1} x(n)e^{\frac{-j2\pi(2n-1)k}{2N}}
$$
(14)

where $2n-1$ represents the circular odd integers from $0 \rightarrow N-1$. Keep in mind, that $k \in \{0, 2N-1\}$, so the second equation will be circularly evaluated twice So that

$$
Y(k) = e^{\frac{j2\pi k}{2N}} \sum_{n=0}^{N-1} x(n) e^{\frac{-j2\pi nk}{N}} = e^{\frac{j2\pi k}{2N}} X((k))_N, \quad k \in \{0, 2N-1\}
$$
 (15)

(b) Given

$$
X_3[k] = \frac{1}{N} \sum_{l=0}^{N-1} X_1[l] X_2[(k-l))_N]
$$

show that $x_3[n] = x_1[n]x_2[n]$

First, we know from sampling as well as the convolution theorem, that

$$
N\sum_{r=-\infty}^{\infty} \delta[n-rN] \longleftrightarrow \sum_{k=-\infty}^{\infty} \delta(f-k/N)
$$
 (16)

so that

$$
\frac{1}{N} \sum_{l=0}^{N-1} X_1[l] X_2[(k-l))_N] = \sum_{l=0}^{N-1} \sum_{n=0}^{N-1} \left(x_1[n] e^{\frac{-j2\pi n l}{N}} x_2[n] e^{\frac{-j2\pi n (k-l)}{N}} \right) \tag{17}
$$

$$
\sum_{n=0}^{N-1} x_3[n]e^{\frac{-j2\pi nk}{N}} = \sum_{l=0}^{N-1} \sum_{n=0}^{N-1} \left(x_1[n]x_2[n]e^{\frac{-j2\pi nl}{N}}e^{\frac{j2\pi nl}{N}}e^{\frac{-j2\pi nk}{N}} \right)
$$
(18)

$$
\sum_{n=0}^{N-1} x_3[n]e^{\frac{-j2\pi nk}{N}} = \sum_{n=0}^{N-1} x_1[n]x_2[n]e^{\frac{-j2\pi nk}{N}}
$$
(19)

and then assuming element-wise equivalence, it is shown that

$$
x_3[n] = x_1[n]x_2[n], \quad n = 0, 1, \dots, N - 1 \tag{20}
$$

4. Consider an FIR lattice filter with coefficients $K_1 = 0.8$, $K_2 = -0.47$, $K_3 = 0.75$

(a) Draw the FIR lattice filter

(b) The difference equations are

$$
y(n) = f_2(n) + 0.75g_2(n-1)
$$
\n(21)

$$
g_2(n) = g_1(n-1) - 0.47f_1(n)
$$
\n(22)

$$
f_2(n) = f_1(n) - 0.47g_1(n-1)
$$
\n(23)

$$
g_1(n) = x(n-1) + 0.8x(n)
$$
\n(24)

$$
f_1(n) = x(n) + 0.8x(n-1)
$$
\n(25)

which gets reduced down to

$$
y(n) = x(n) + 0.0715x(n-1) - 0.152x(n-2) + 0.75x(n-3)
$$
 (26)

(c) Draw the equivalent direct-form structure

5. Given

$$
H(z) = 0.1325 - 0.0867z^{-1} + 0.4205z^{-2} + 1.3592z^{-3} + 0.4205z^{-4} - 0.0867z^{-5} + 0.1325z^{-6}
$$

(a) We can split up the transfer function into multiple sections by knowing its roots. Since all roots are complex, the transfer function $H(z)$ can be split up into multiple 2nd-order sections

$$
H(z) = \frac{1}{0.1325} S_1(z) S_2(z) S_3(z)
$$
\n(27)

where

$$
S_1(z) = z^2 - 1.0184z + 31.0874
$$
\n(28)

$$
S_2(z) = z^2 + 0.3968z + 1\tag{29}
$$

$$
S_3(z) = z^2 - 0.0328z + 0.0322\tag{30}
$$

Where we can take any sections $S_k(z)$ and $S_l(z)$ where $k \neq l$ to form either the first or second part of the cascaded filter. For this example, we will create the first filter F_1 out of $S_1(z)$ and $S_2(z)$ while the second filter F_2 will be $S_3(z)$. Using the algorithm in class, we find (by hand) that

$$
F_1: K_1 = 0.1984, K_2 = 1, K_3 = -0.0317, K_4 = 31.0874
$$
 (31)

 $F_2: K_1 = -0.0317, K_2 = 0.0322$ (32)

We can confirm this in MATLAB, where I wrote a function to do just this

```
1 function [Ks, gain] = fir_l lattice(Hz)2 % Author: Arpad Voros
3 % fir lattice() routine takes in coefficients of a transfer function
4 % and determines the FIR Lattice coefficients
5 % INPUT: Hz − array of H(z) coefficients
6 % OUTPUT: Ks − coefficents of lattice, ordered from input to ...
      output cell array, if needs to be split up (special cases)
7 % gain − the amount of gain for a given lattice
8
9 % initialize some variables
10 order = length(Hz) - 1;
11
12 % initialize outputs
13 Ks = \{\}\;14 gain = {};
15
16 % check special case
17 if Hz(1) == Hz(order + 1)18 % include first gain
19 if Hz(1) \neq 120 \qquad \qquad gain{end + 1} = Hz(1);
21 end
22
23 % split the roots
24 rootsHz = roots(Hz);
25 split = ceil(length(rootsHz) / 2);
26 if rootsHz(split) == conj(rootsHz(split + 1))
27 split = split + 1;
28 end
29
30 % recursive call
31 [Ks{end + 1}, gain{end + 1}] = fir.lattice(poly(rootsHz(1:split)));
32 [Ks{end + 1}, gain{end + 1}] = fir.lattice(poly(rootsHz(split + ...
          1:end)));
33 else
34 % normalize, place into Az
35 if Hz(1) \neq 136 \qquad \qquad gain\{end + 1\} = Hz(1);37 Az = Hz / Hz(1);
38 else
39 Az = Hz;40 end
41
42 % temporary coefficients to be appended to Ks
43 K = zeros(order, 1);
44
45 % loop
46 for n = order:−1:3
47 % get last coefficient
```

```
48 K(n) = Az(n + 1);49
50 % reverse coefficients
51 Bz = flip(Az);
52
53 % get new value of Az
54 Az = (Az - K(n) * Bz) / (1 - K(n)^2);<br>55 Az = Az(1:end - 1);Az = Az(1:end - 1);56 end
57 % final two coefficients
58 K(2) = Az(3);
59 K(1) = Az(2) / (1 + K(2));60
61 % append to rest of coefficients
62 Ks{end + 1} = K;
63 end
64
65 % delete empty cells
66 gain = gain(¬cellfun('isempty', gain));
67 end
```
So by calling fir_lattice in the script below, we get

```
1 % coefficients of transfer function
2 h = [0.1325, −0.0867, 4.205, 1.3592, 4.205, −0.0867, 0.1325];
3
4 % display gain and K coefficients
5 [result, gain] = fir_lattice(h);
6 celldisp(gain);
7 celldisp(result);
gain{1} = result{1}{1} = result{2}{1} =0.1325 0.1984 -0.0317
                      1.0000 0.0322
                      -0.0317
                     31.0874
```
(b) Superscripts in this section do not indicate exponentiation, but rather help distinguish between F_1 and F_2 , where f_k^1 and g_k^1 correspond to F_1 while f_k^2 and g_k^2 correspond to F_2 . The appropriate difference equations for the cascaded filter above are

$$
y(n) = f_1^2(n) + 0.0322g_1^2(n-1)
$$
\n(33)

$$
g_1^2(n) = f_4^1(n-1) - 0.0317 f_4^1(n)
$$
\n(34)

$$
f_1^2(n) = f_4^1(n) - 0.0317 f_4^1(n-1)
$$
\n(35)

$$
f_4^1(n) = f_3^1(n) + 31.0874g_3^1(n-1)
$$
\n(36)

$$
g_3^1(n) = g_2^1(n-1) - 0.0317 f_2^1(n)
$$
\n(37)

- $f_3^1(n) = f_2^1(n) 0.0317g_2^1(n-1)$ (38)
	- $g_2^1(n) = g_1^1(n-1) + f_1^1$ (39)
	- $f_2^1(n) = f_1^1(n) + g_1^1(n-1)$ (40)
- $g_1^1(n) = x(n-1) + 0.1984x(n)$ (41)
- $f_1^1(n) = x(n) + 0.1984x(n-1)$ (42)